A close-up photograph of a microelectronic assembly process. A fine wire is being positioned by a mechanical arm over a small component on a circuit board. The background is blurred, showing other parts of the assembly station.

RF Wild - Measurements, Modeling, Microelectronics



RF Microelectronics – Signals:
properties, simulation & measurements

Outline



- Introduction to signals:
 - Properties
 - Units
- Simulation of signals :
 - Time-domain
 - Frequency domain
- Introduction to measurements:
 - Signal generator, spectrum analyzer and oscilloscope
 - Time and frequency domain measurements

Introduction to signals



- Let's begin by the "simple" case of a sine:

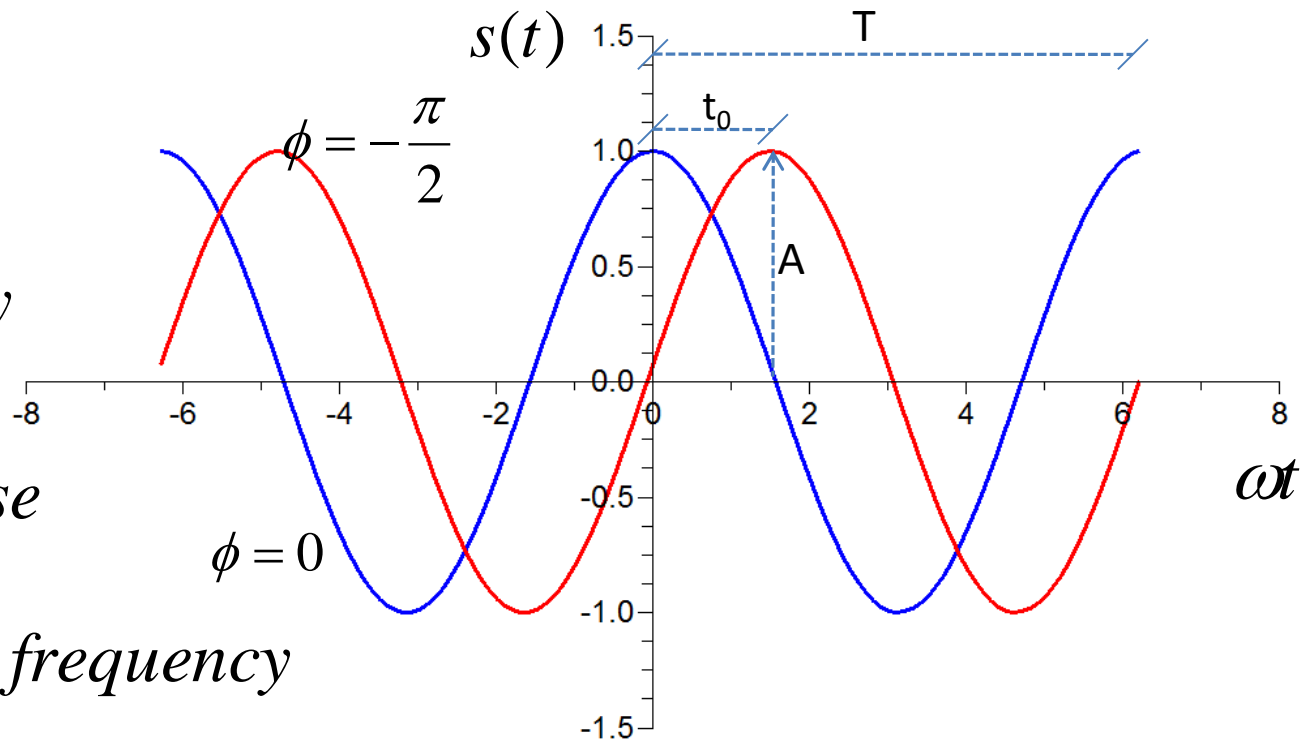
$$s(t) = A \cos(\omega t + \phi)$$

$A = \text{amplitude}$

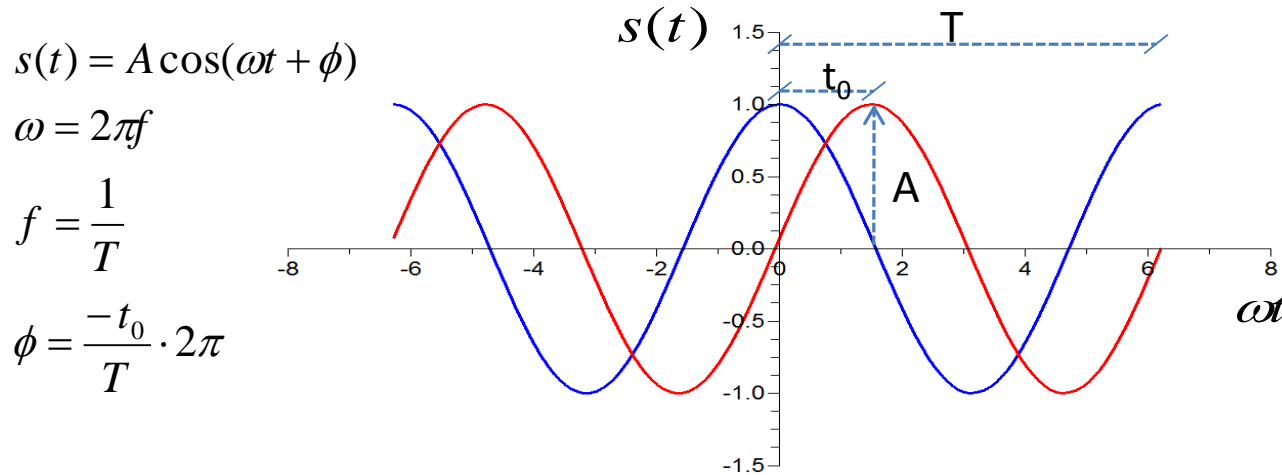
$$f = \frac{1}{T} = \text{frequency}$$

$$\phi = \frac{-t_0}{T} \cdot 2\pi = \text{phase}$$

$$\omega = 2\pi f = \text{angular frequency}$$



Introduction to signals



In this case, we say the **red curve** is **delayed** with respect to the **blue curve**, (or the **blue curve** is **in advance** with respect to the **blue curve**).

- Determine the expression of $s(t)$ if:

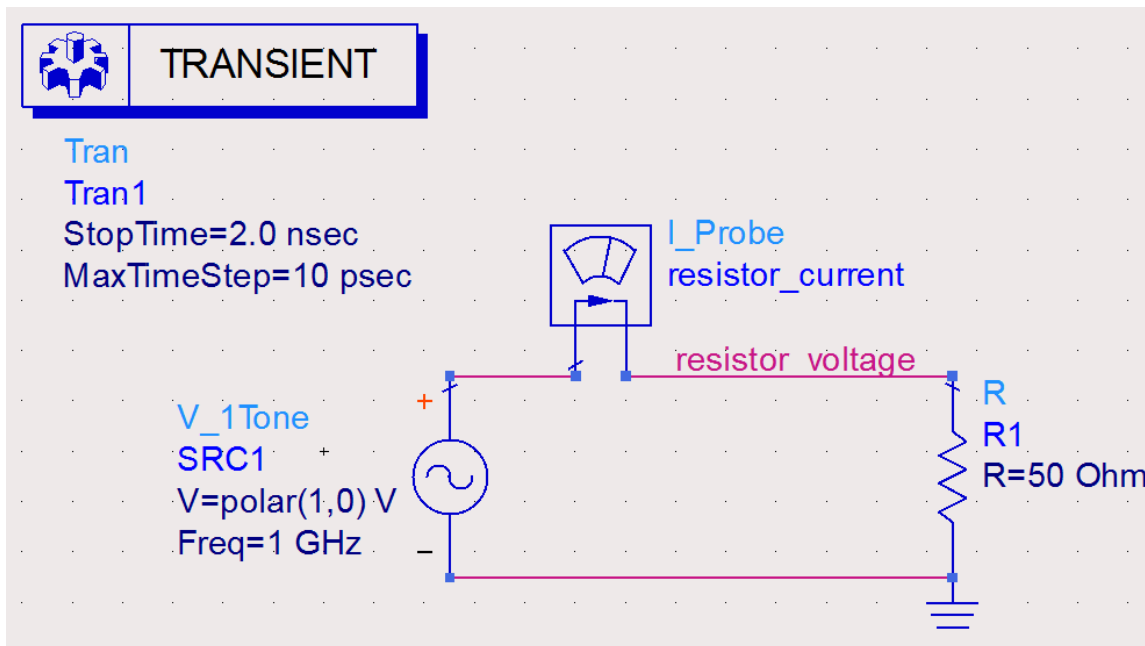
$$A = 1; T = 10^{-9} \text{ s}; t_0 = -0.25 \text{ ns}$$

$$A = 2; T = 10^{-10} \text{ s}; t_0 = 0.5 \cdot 10^{-10} \text{ s}$$

Introduction to signals



- What happens if we apply a 1V, 1GHz voltage signal to a (50 Ω) resistor?

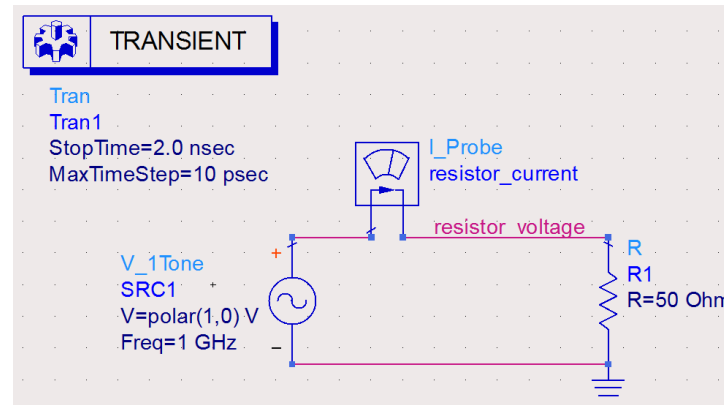


Try to calculate
“instantaneous”
current and power

Introduction to signals



- What happens if we apply a 1V, 1GHz voltage signal to a (50 Ω) resistor?



$$resistor_voltage(t) = 1 \cdot \cos(2\pi 10^9 t)$$

$$resistor_current(t) = \frac{1}{50} \cdot \cos(2\pi 10^9 t)$$

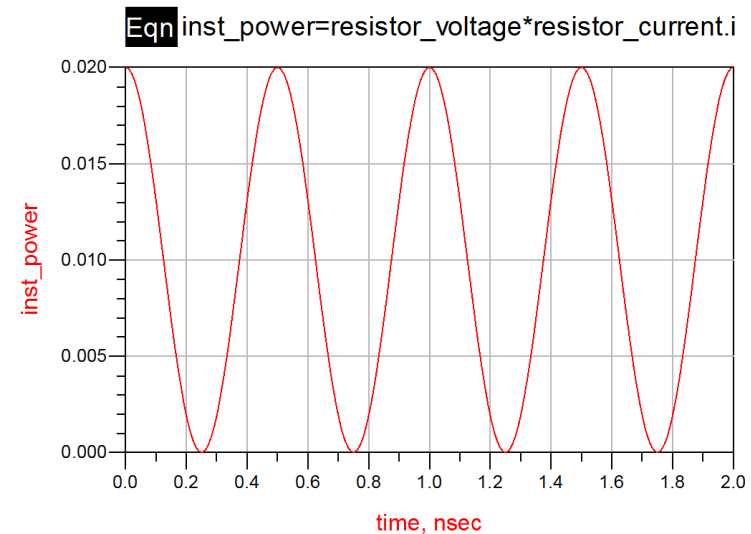
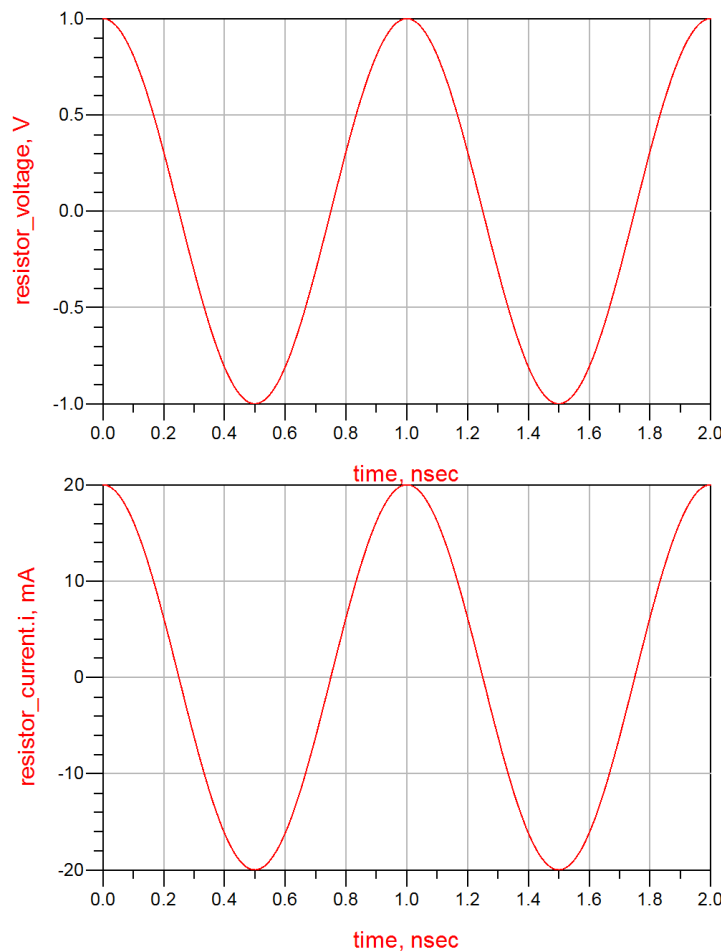
$$inst_power(t) = 1 \cdot \cos(2\pi 10^9 t) \cdot \frac{1}{50} \cdot \cos(2\pi 10^9 t) \Rightarrow$$

$$inst_power(t) = \frac{1^2}{2 \cdot 50} + \frac{1^2}{2 \cdot 50} \cdot \cos(2 \cdot 2\pi 10^9 t) \quad \left\{ \begin{array}{l} P_{DC} = 0.01W \\ P_{f=2GHz} = 0.01W_{peak} \end{array} \right.$$

Introduction to signals



- Look at the simulation results:



$$\text{resistor_voltage}(t) = 1 \cdot \cos(2\pi 10^9 t)$$

$$\text{resistor_current}(t) = \frac{1}{50} \cdot \cos(2\pi 10^9 t)$$

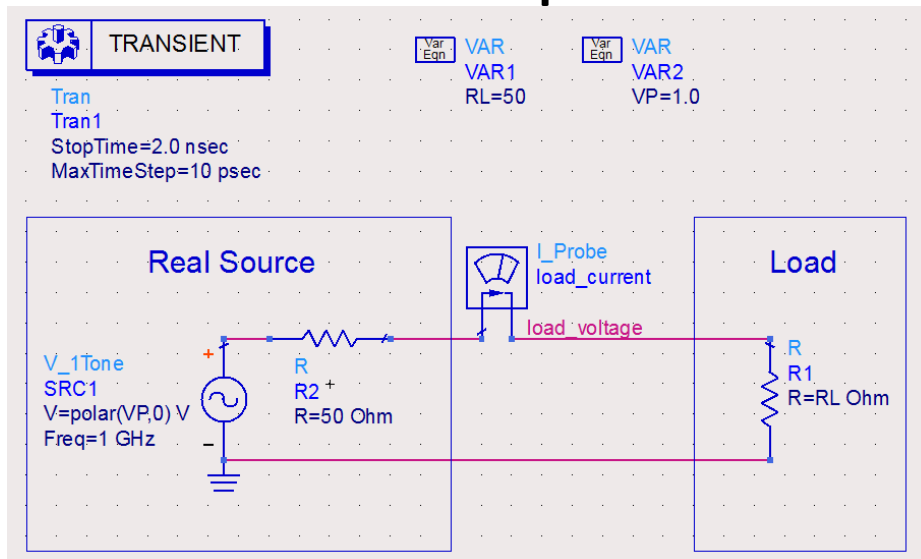
$$\text{inst_power}(t) = 0.01W + 0.01W \cdot \cos(2 \cdot 2\pi 10^9 t)$$

Sounds good?

Introduction to signals



- As you know, ideal voltage (or current) sources do not exist: Real sources have internal impedances!
- We will consider RF sources with a nominal 50Ω internal impedance:

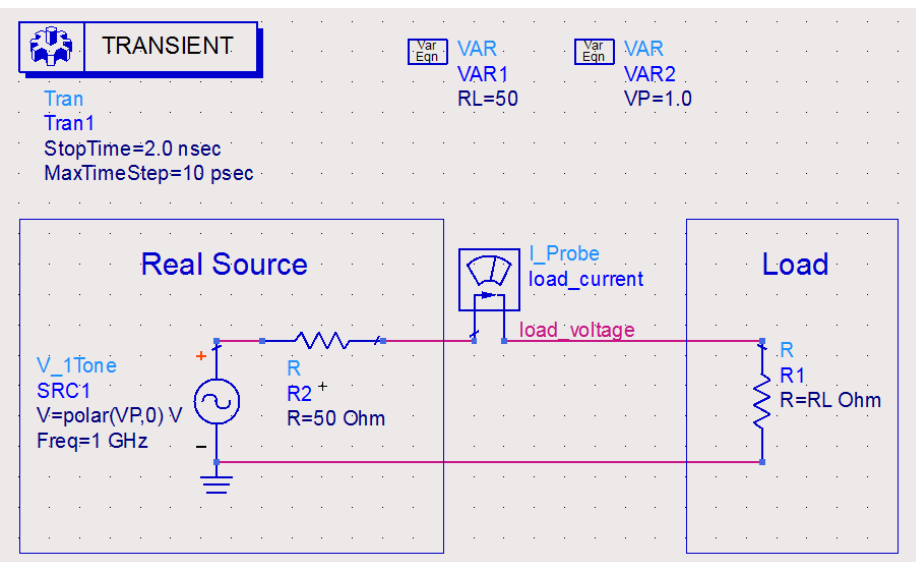


Let's do some math again: try to calculate the DC (mean) **available** power of the source

Introduction to signals



- The available power of the source corresponds to the power delivered when the load is the complex conjugate of the internal impedance (more on this later). In this case, if $R_L = 50\Omega$!



$$load_voltage(t) = \frac{V_P}{2} \cdot \cos(2\pi 10^9 t)$$

$$load_current(t) = \frac{V_P}{2 \cdot 50\Omega} \cdot \cos(2\pi 10^9 t)$$

$$load_power(t) = \frac{V_P^2}{8 \cdot 50\Omega} + \frac{V_P^2}{8 \cdot 50\Omega} \cdot \cos(2 \cdot 2\pi 10^9 t)$$

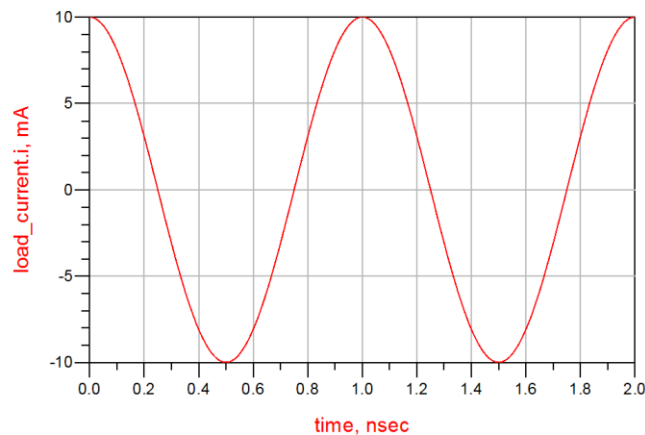
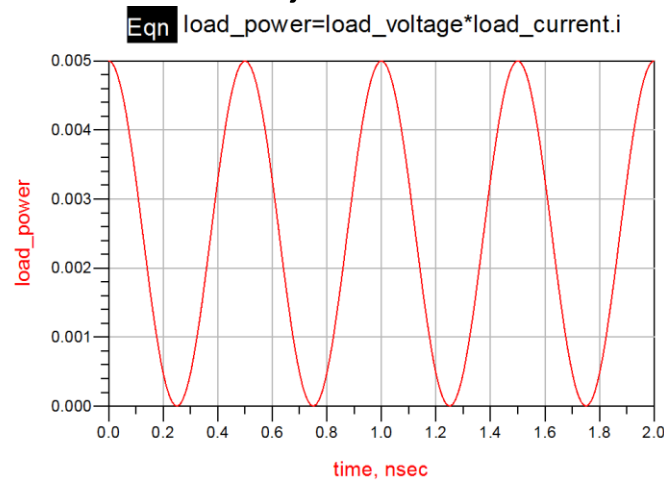
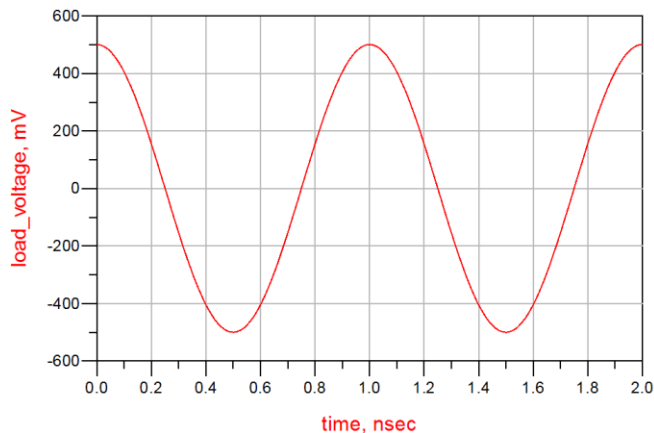


Available power
(load independent!)

Introduction to signals



- Check the results for $V_P=1V$, $R_L=50\ \Omega$:



$$\text{load_voltage}(t) = \frac{1}{2} \cdot \cos(2\pi 10^9 t)$$

$$\text{load_current}(t) = \frac{1}{2 \cdot 50\Omega} \cdot \cos(2\pi 10^9 t)$$

$$\text{load_power}(t) = \frac{1^2}{8 \cdot 50\Omega} + \frac{1^2}{8 \cdot 50\Omega} \cdot \cos(2 \cdot 2\pi 10^9 t)$$

Available power=2.5 mW

Introduction to signals



- Let's TALK ABOUT POWER!!
- Power in physical systems may vary by **several orders of magnitude**.
 - Human ear may detect a huge range of sound (power) levels
 - Portable phones are able to detect signals from about 10^{-13} W (yes, 0.1 pW!!!) to some mW
- We should use a logarithmic scale: **Decibels!!**

Introduction to signals



- But what is a **Decibel**?
 - It is a logarithmic “unit” which represents a relation (a priori of power) with respect to a given reference.
 - It is very usefull to represent physical variables which vary a lot:

$$Gain(dB) = 10 \log\left(\frac{P}{P_{REF}}\right)$$

Remember:
Gain is unitless!

$$Gain(dB) = 10 \log\left(\frac{P}{P_{REF}}\right) = 10 \log\left(\frac{\frac{V^2}{R}}{\frac{V_{REF}^2}{R}}\right) = 20 \log\left(\frac{V}{V_{REF}}\right)$$

Introduction to signals



- Let's do some math:

$$Gain(dB) = 10\log\left(\frac{P}{P_{REF}}\right) \quad Gain(dB) = 20\log\left(\frac{V}{V_{REF}}\right)$$

- What is the gain (in dB) of an amplifier having an input power of 1 mW and output power of 1 W? **Answer: 30 dB**
- What is the (power) loss in an attenuator whose input voltage is 1 V (peak) and output voltage is 10 mV (peak)?
Answer: 40 dB (-40 dB)

Introduction to signals



- With **Decibels** in mind, we may know think about **power units**!! Units commonly used:
 - dBm (dB scale with respect to 1 mW)
 - dBW (dB scale with respect to 1 W)

$$0dBW = 1W$$

$$10dBm = 10mW = -20dBW$$

$$0dBm = 1mW$$

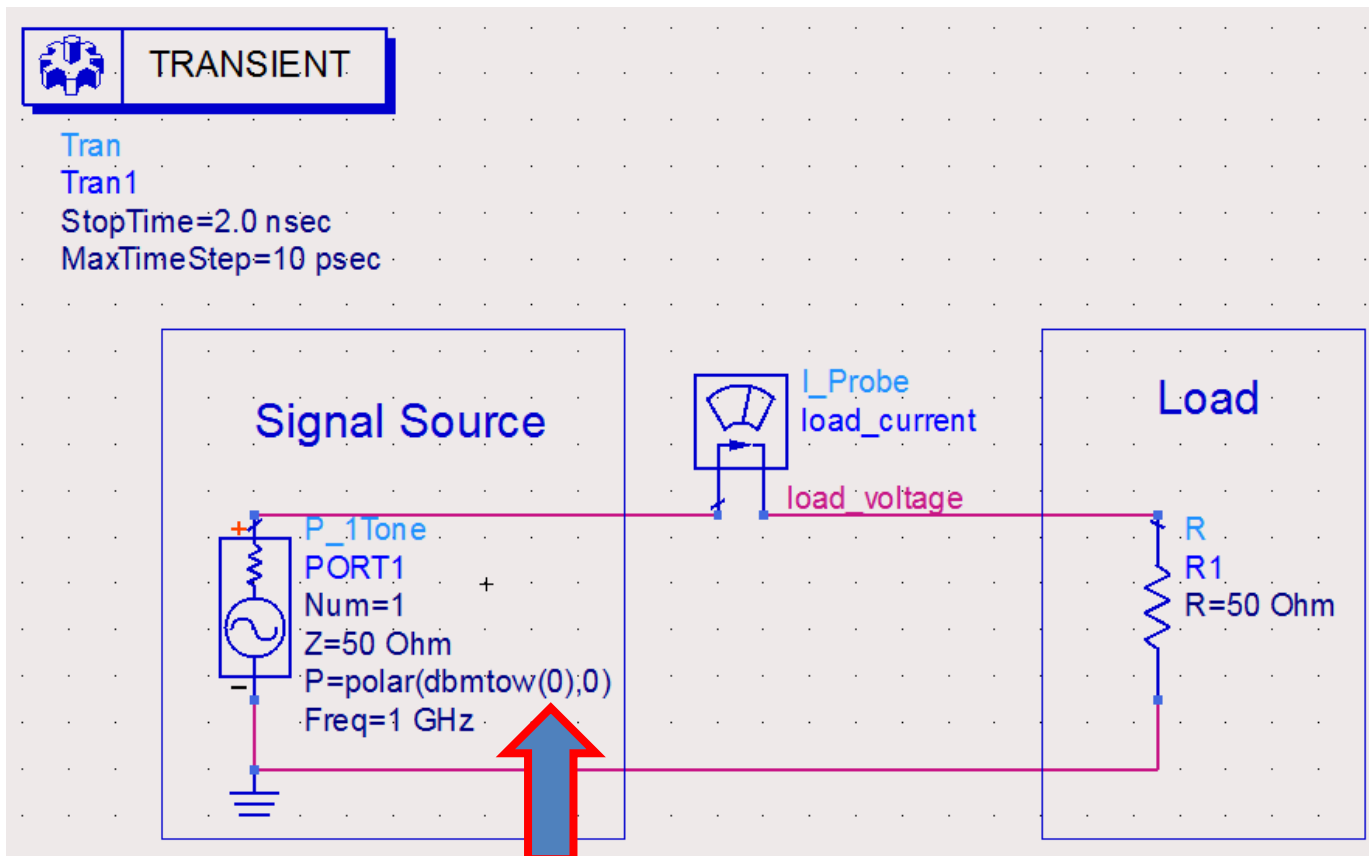
$$-20dBm = 10\mu W = -50dBW$$

Short-range wireless transceivers generally transmit power levels around 0 dBm (1 mW) and are able to receive signals with a power level down to about -100dBm (10^{-13} W). In this latter case, the signal in the (50 Ω) antenna has $2.2 \mu V_{RMS}$!!

Introduction to signals



- Let's use hereafter the signal source below.



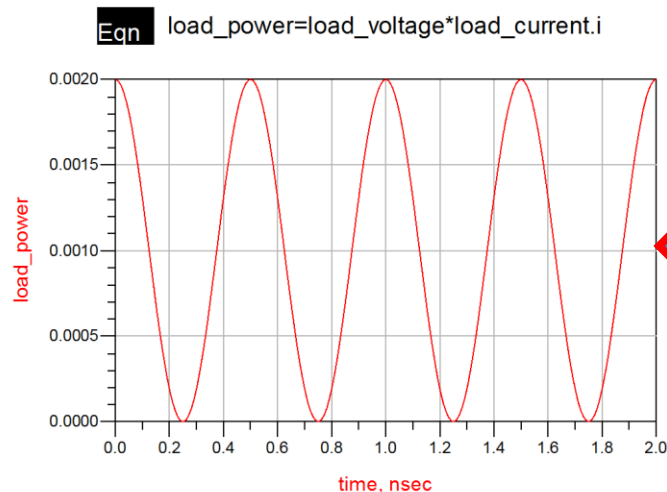
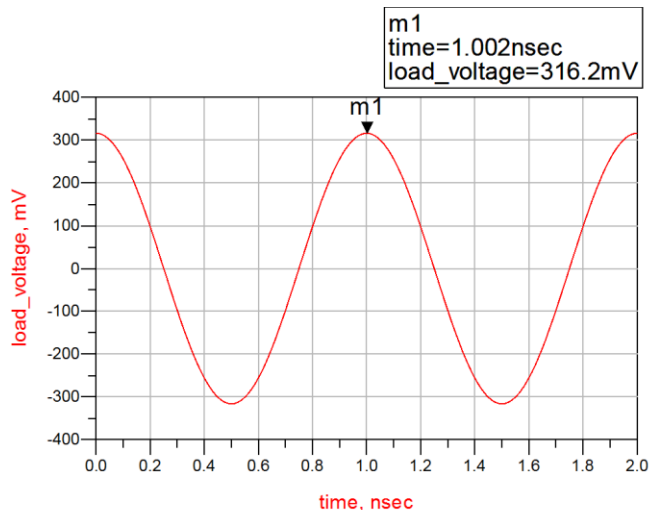
Power in dBm

Calculate
peak voltage
and current
in the load

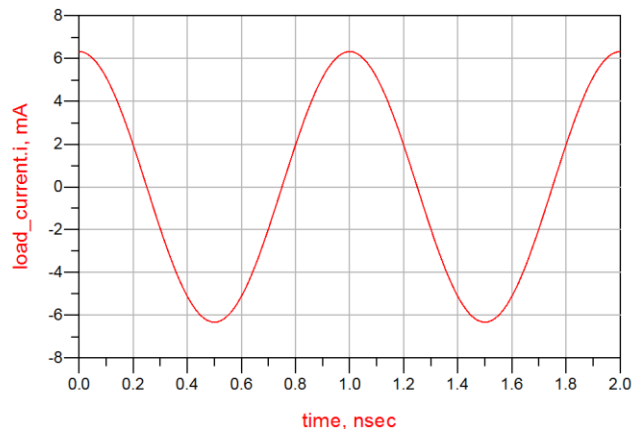
Introduction to signals



- Check the numbers:



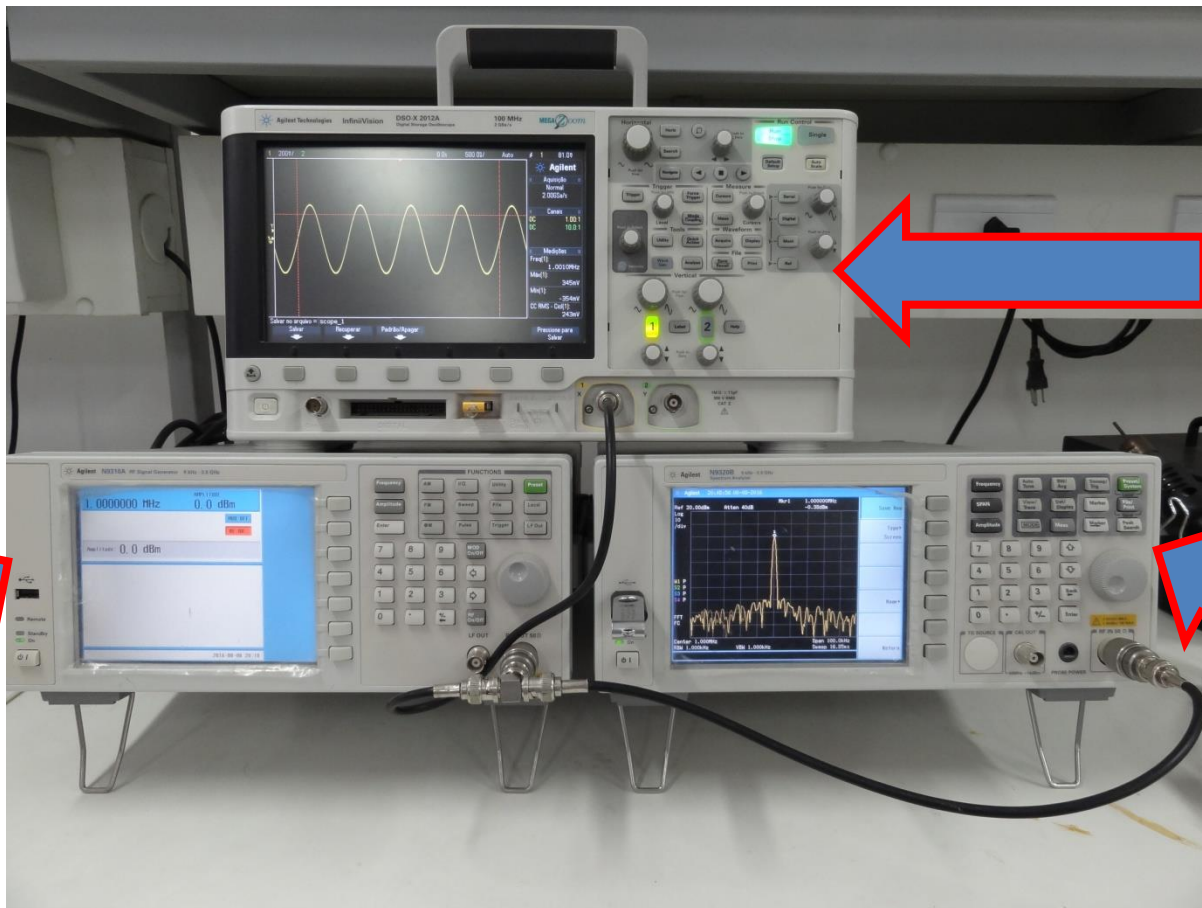
1 mW
(0 dBm)



Introduction to measurements



- How does an RF signal source look like?



Signal
source

Oscilloscope

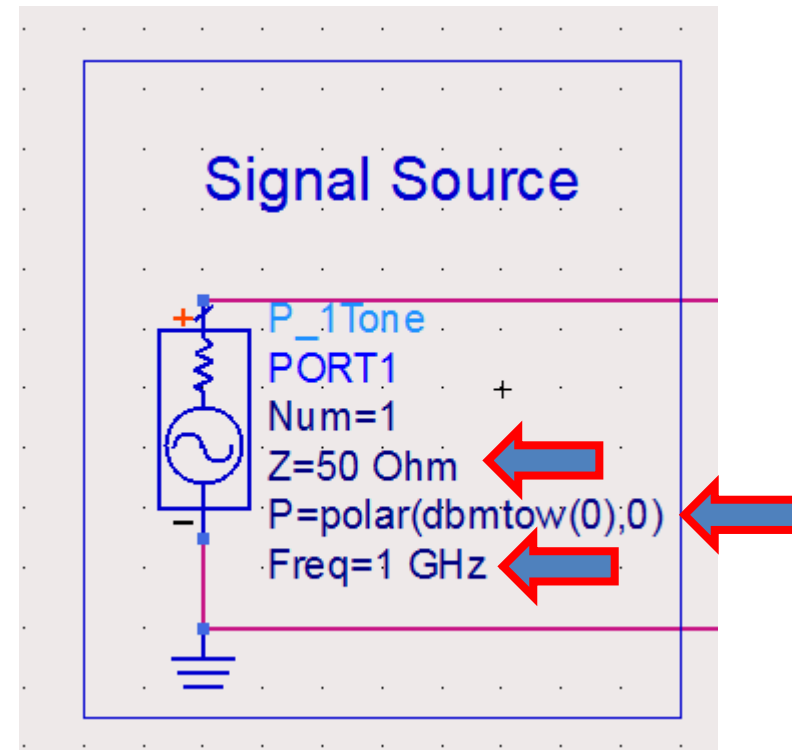
Spectrum
analyzer

Introduction to measurements



- Compare with our simulation model:

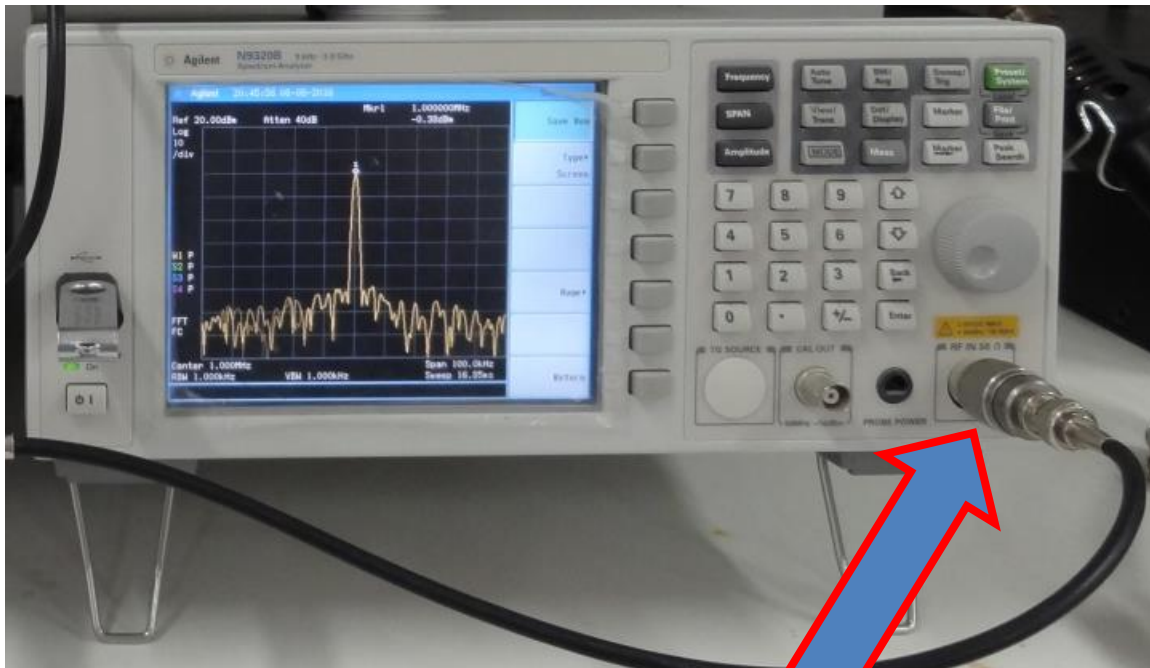
Frequency Power



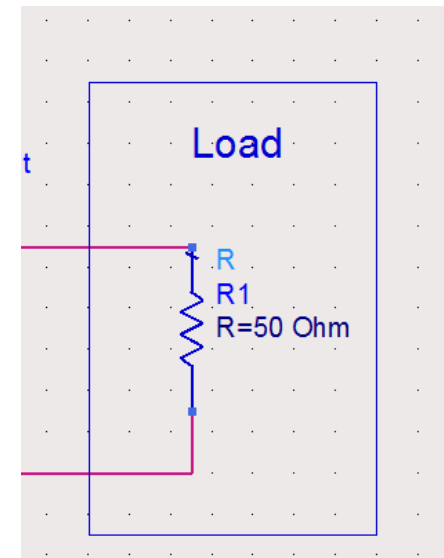
Introduction to measurements



- What about our spectrum analyzer?



RF in
 $50\ \Omega$

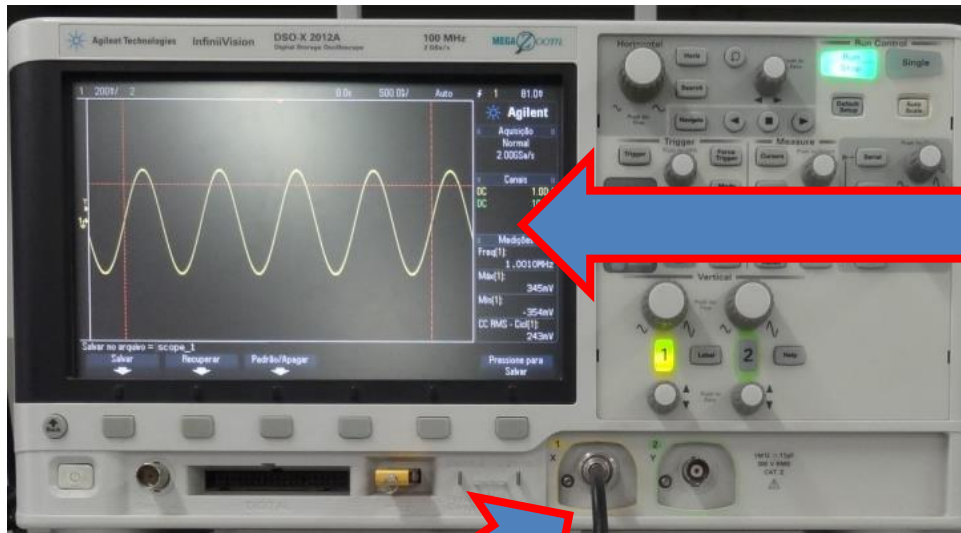


It works as a
(nominally) $50\ \Omega$ load!

Introduction to measurements



- And what about the oscilloscope???



We will use the oscilloscope to see the signals in the time domain

$1\text{ M}\Omega // 11\text{ pF}$

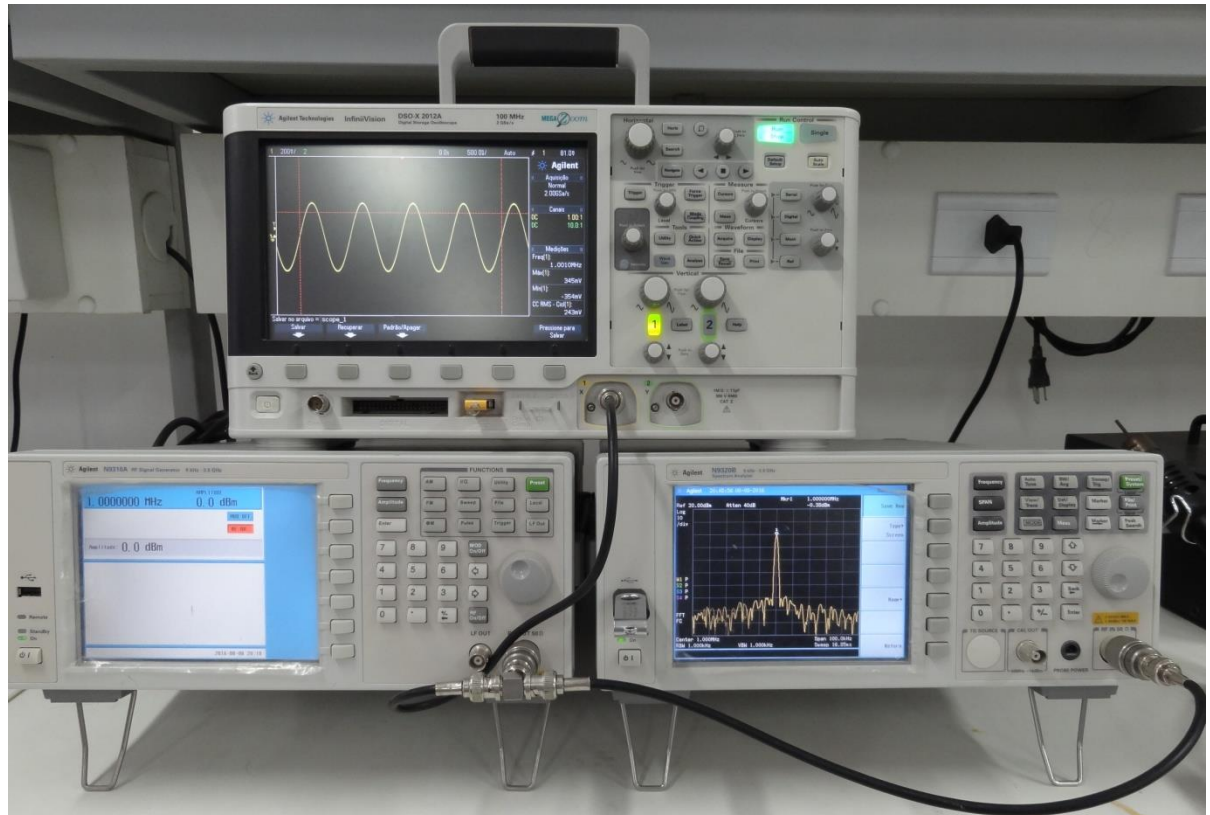
Until 10 MHz, its impedance is about 30 x greater than 50Ω (open circuit!)

Introduction to measurements



- Are you ready for the **FUN**??? Let's **MEASURE**!

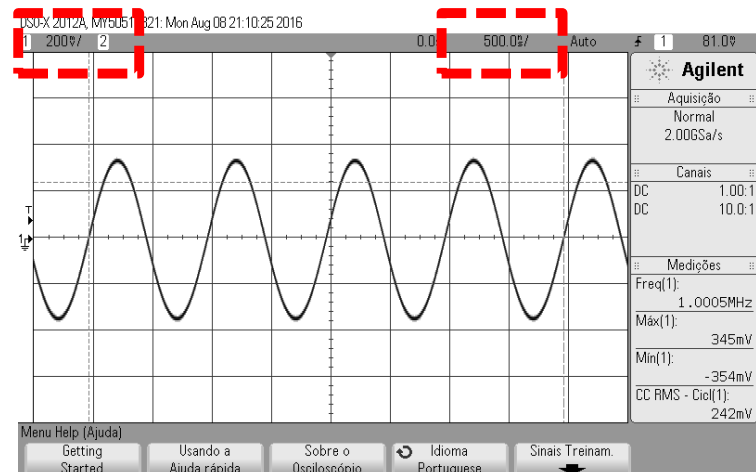
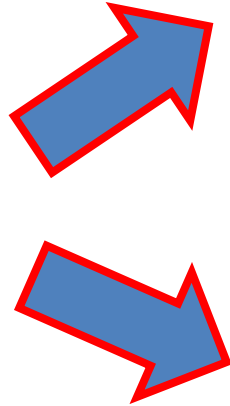
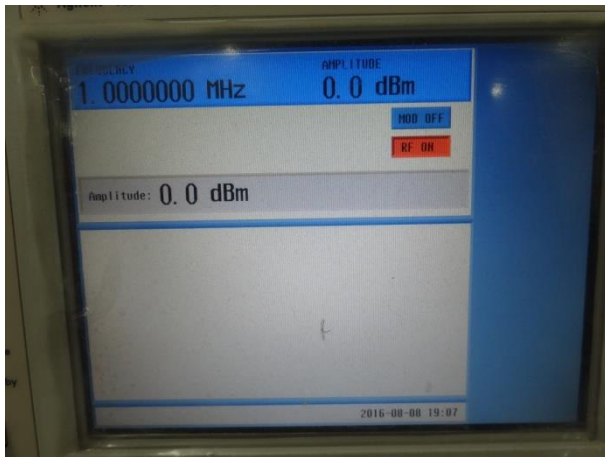
We will use
this setup:



Introduction to measurements



- Set the source to **0 dBm**, **1 MHz**:

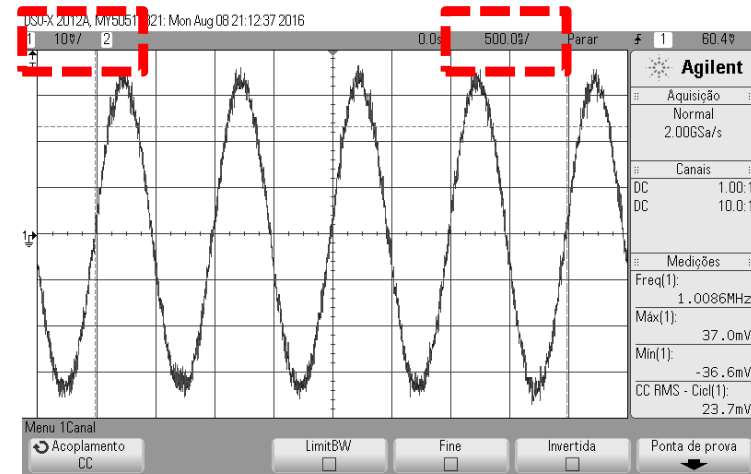
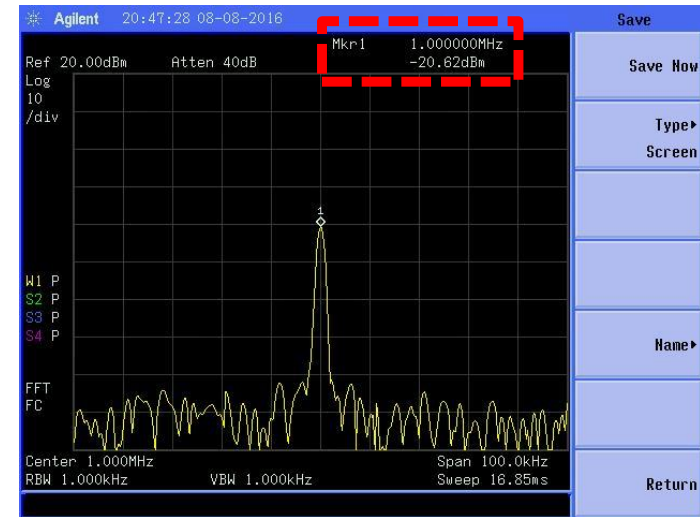
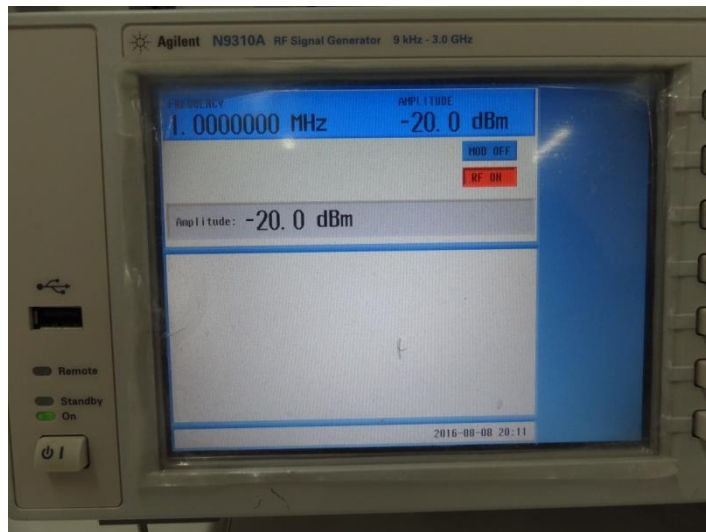


Check the numbers!!
(look closely😊)

Introduction to measurements



- Set the source to **-20 dBm**, **1 MHz**:

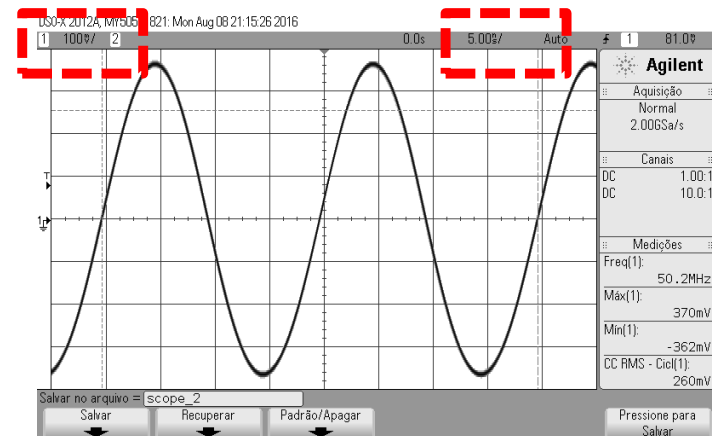
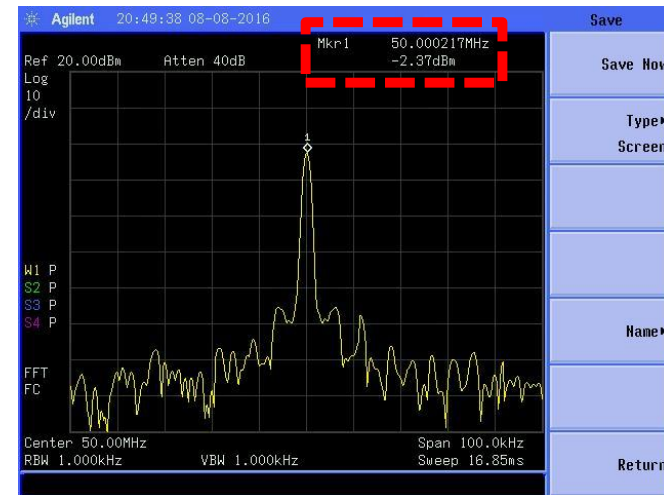
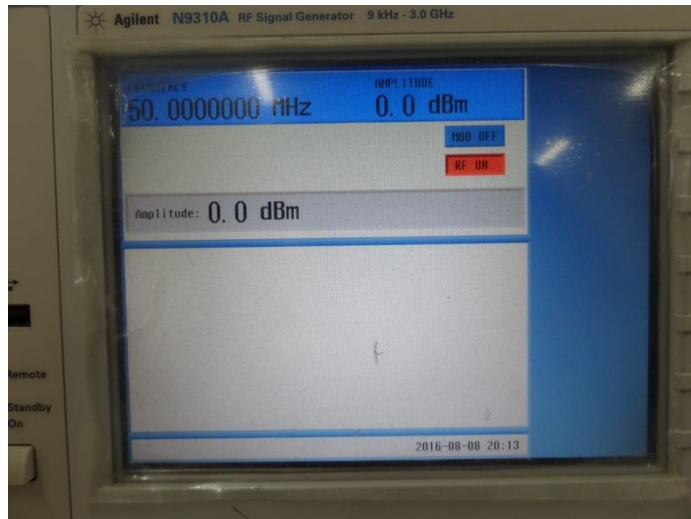


Check the numbers!!
(look closely😊)

Introduction to measurements



- Set the source to **0 dBm**, **50 MHz**:



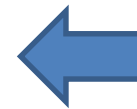
Check the numbers!!
(look closely😊)

Introduction to modulated signals



- Relation between wavelength (λ) and the signal's frequency (f):

$$\lambda = \frac{c}{f} = \frac{3 \cdot 10^8 \text{ m/s}}{f}$$



Propagation speed
in the medium
considered

- To obtain a “satisfactory” performance in the transmission of electromagnetic signals, the physical length of the antenna should be on the order of λ (generally $\lambda/4$).

If we decided to transmit audio signals (eg. 10kHz) directly through electromagnetic waves, the antenna should have about 7,5 km!!!

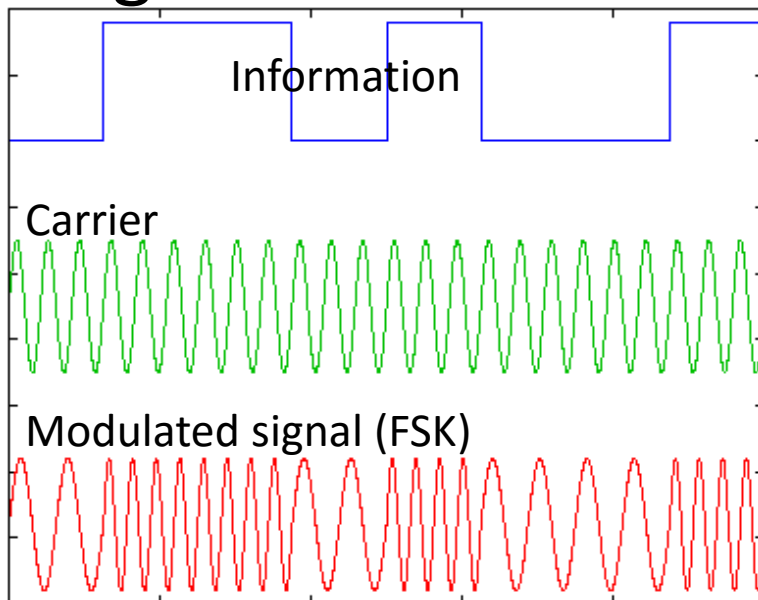
Introduction to modulated signals



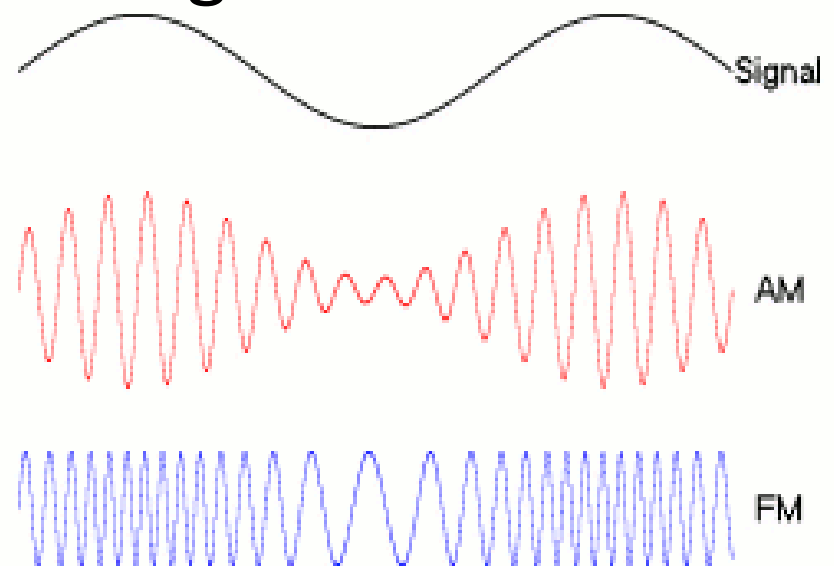
- We deal with that by **modulating** a **carrier**!

Modulate is the process of varying a signal's (carrier) parameter according to the information to be transmitted.

Digital modulation:



Analog modulation:






Introduction to modulated signals



- Let's start by the "simple" case of AM DSB FC created by a single modulating tone:

$$AM_{DSBFC}(t) = [1 + M \cdot \cos(\omega_m t)] \cdot K \cos(\omega_c t)$$

Modulation index Modulating frequency Carrier frequency

If **M=1** (100% modulation) and **K=1** (carrier amplitude):

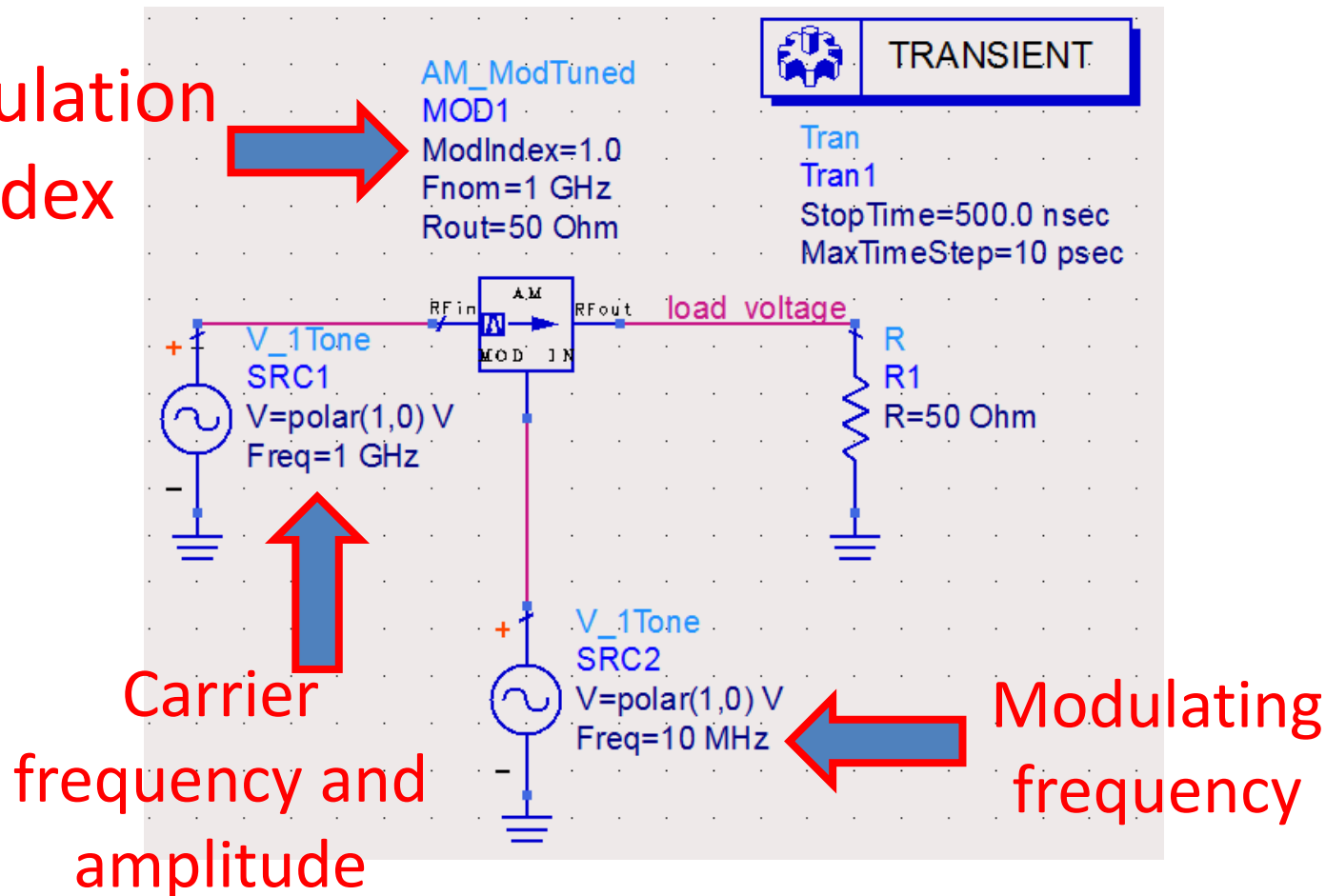
$$AM_{DSBFC}(t) = [1 + \cos(\omega_m t)] \cdot \cos(\omega_c t)$$

Introduction to modulated signals



- Let's simulate that with the setup below:

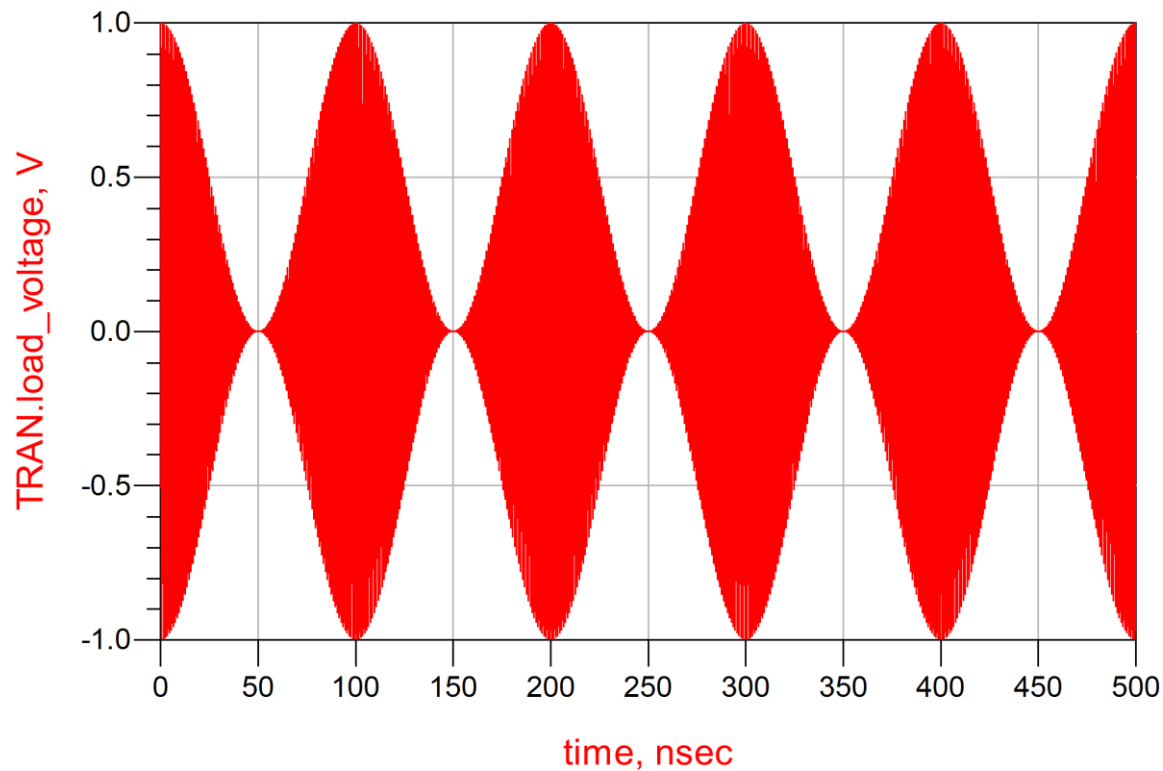
Modulation
index



Introduction to modulated signals



- Look at the results:



$$AM_{DSBFC}(t) = [1 + \cos(\omega_m t)] \cdot \cos(\omega_c t)$$

Introduction to modulated signals



- For a generic carrier amplitude, but as long as $M=1$:

$$AM_{DSBFC}(t) = [1 + \cos(\omega_m t)] \cdot K \cos(\omega_c t) \Rightarrow$$
$$AM_{DSBFC}(t) = \begin{cases} \frac{K}{2} \cos[(\omega_c - \omega_m)t] + \\ K \cos(\omega_c t) + \\ \frac{K}{2} \cos[(\omega_c + \omega_m)t] \end{cases}$$

Lower sideband

Carrier

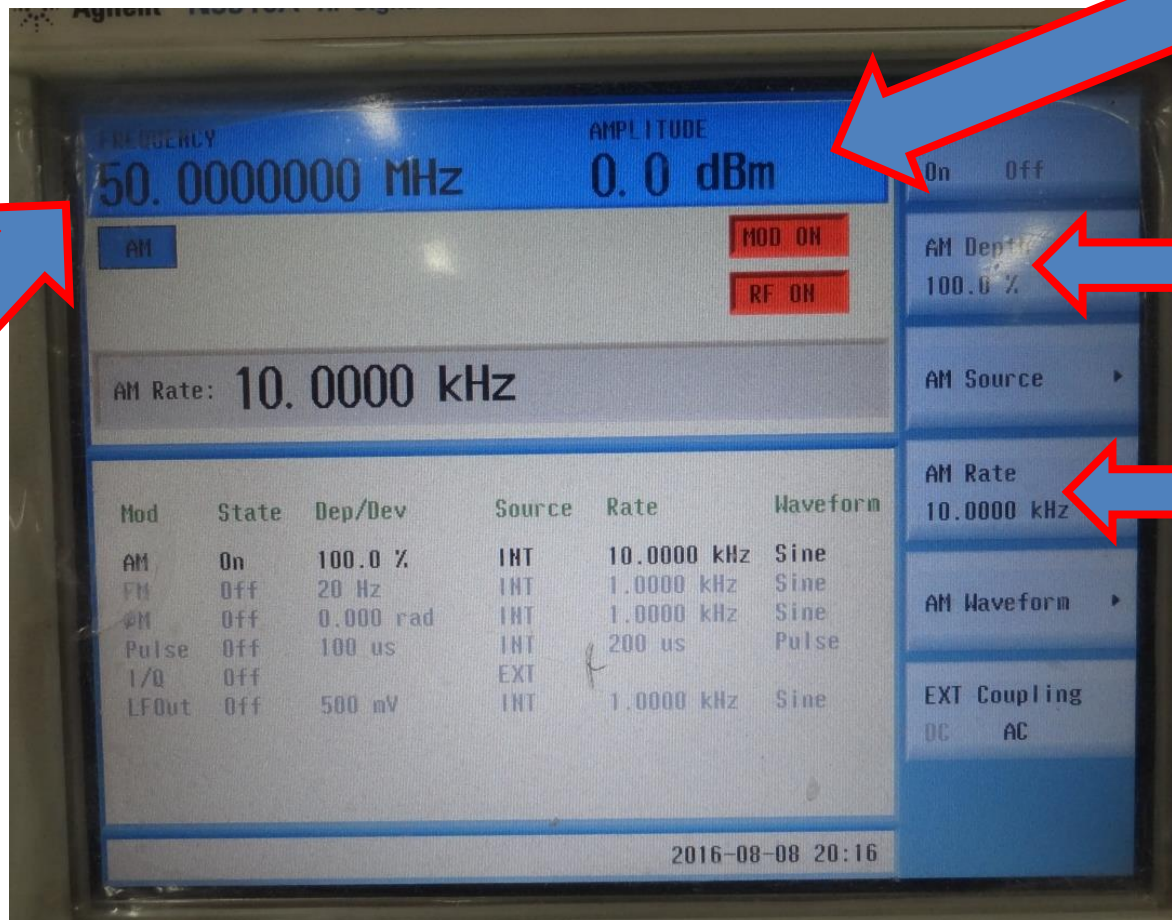
Upper sideband

Notice that for $M=1$, the **sidebands** have **HALF** de **VOLTAGE** of the **carrier**. This means -6 dB!!!

Introduction to measurements



- Let's check that with measurements:



Carrier
power
(0 dBm)

M=1

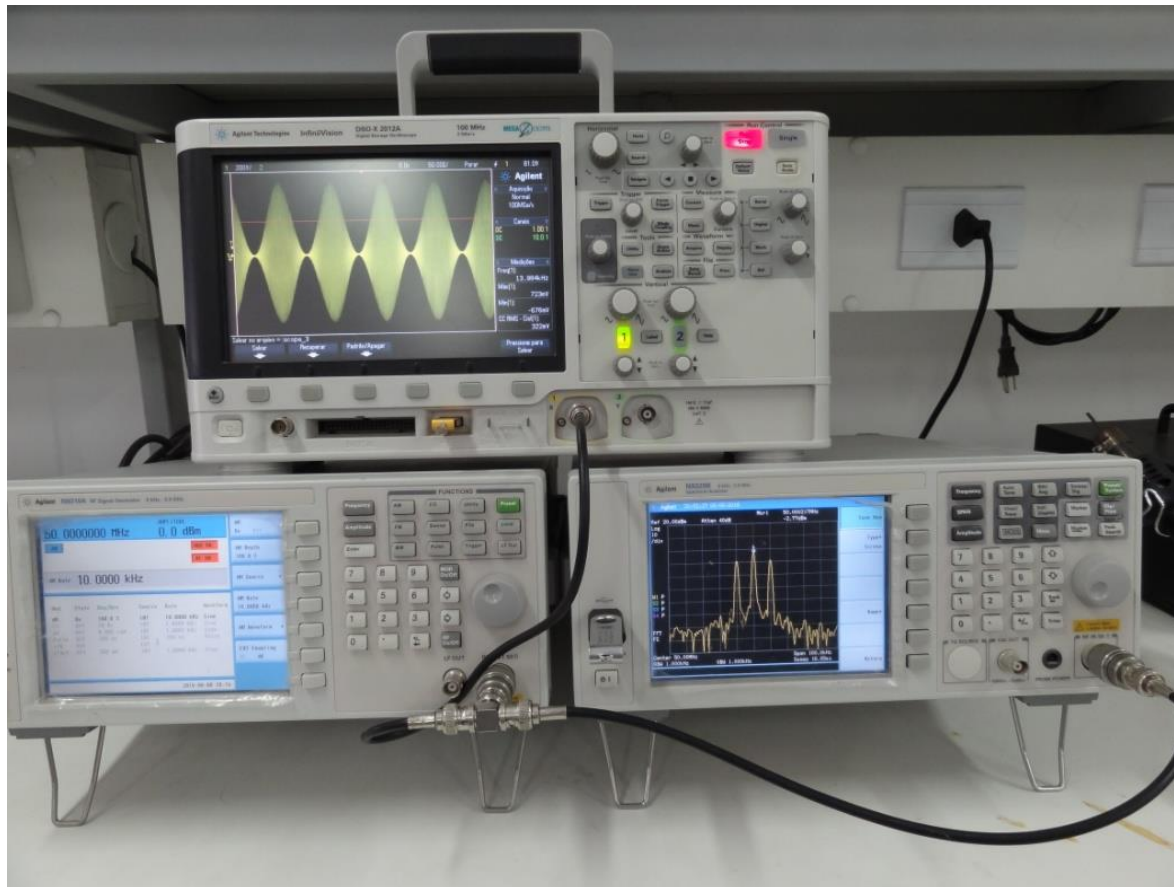
Modulating
frequency
(10 kHz)

Carrier
frequency
(50 MHz)

Introduction to measurements



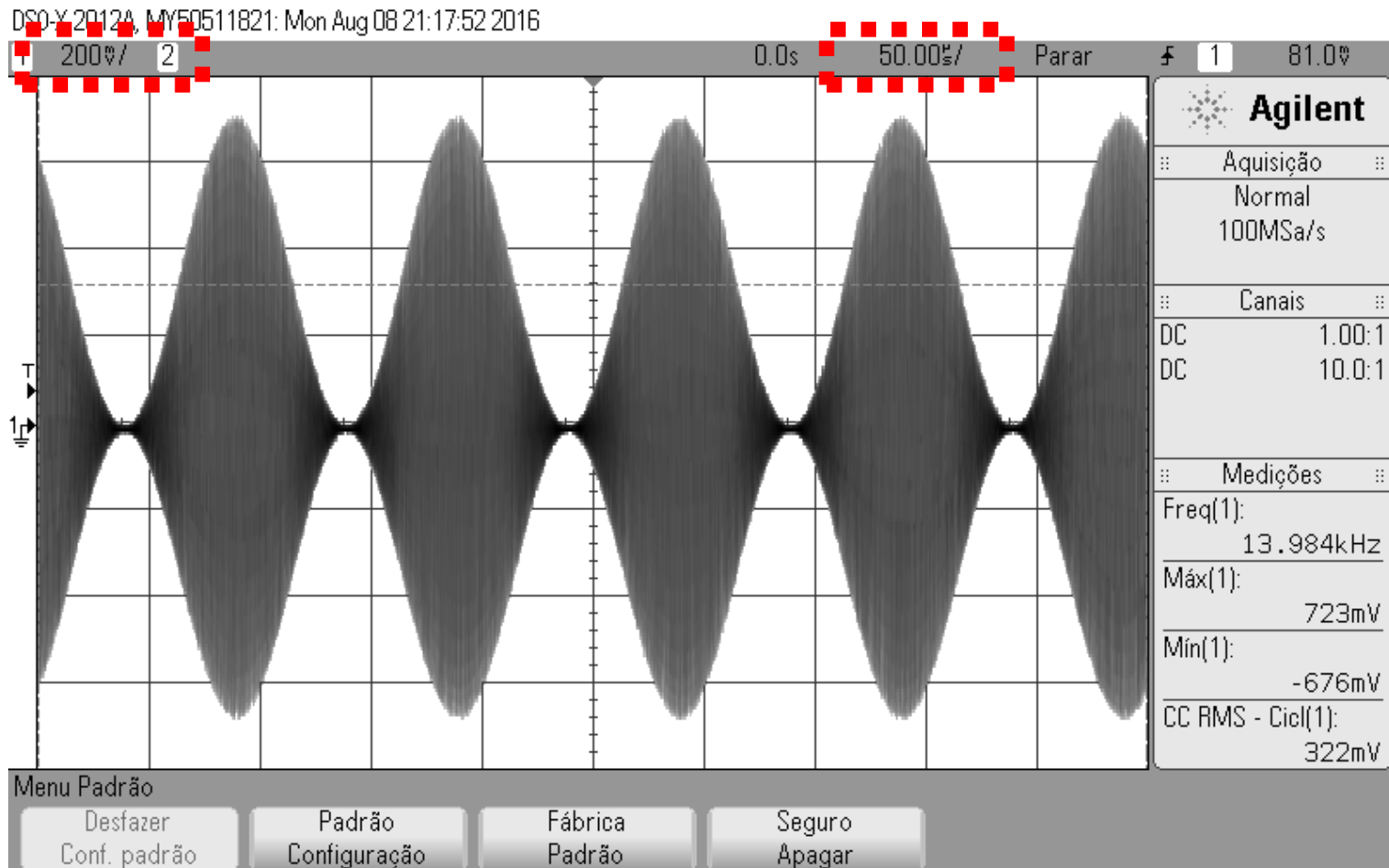
- The setup should look like this:



Introduction to measurements



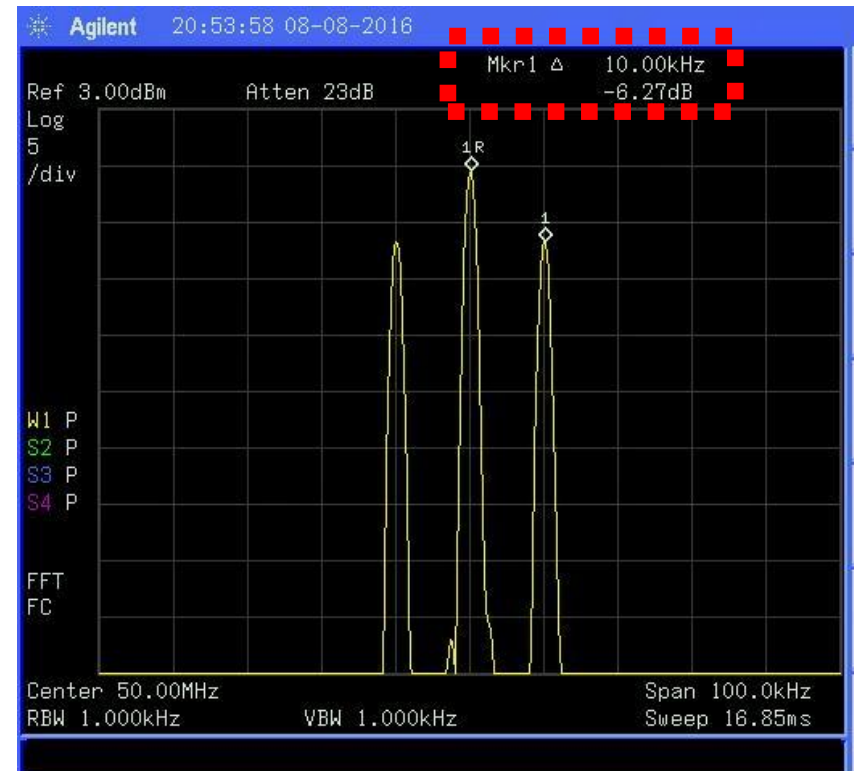
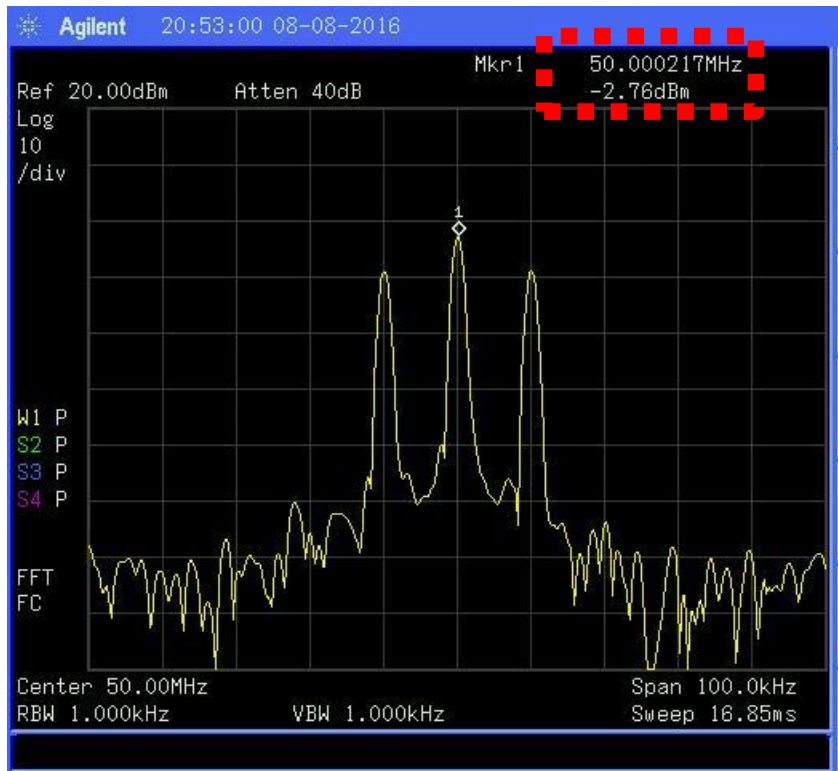
- Check the time-domain waveform:



Introduction to measurements



- Look at the frequency domain measurements:



Sounds good?

Introduction to modulated signals



- What if we change to $M=0.5$?

$$AM_{DSBFC}(t) = [1 + 0.5 \cdot \cos(\omega_m t)] \cdot K \cos(\omega_c t) \Rightarrow =$$

$$AM_{DSBFC}(t) = \begin{cases} \frac{K}{4} \cos[(\omega_c - \omega_m)t] + \\ K \cos(\omega_c t) + \\ \frac{K}{4} \cos[(\omega_c + \omega_m)t] \end{cases}$$

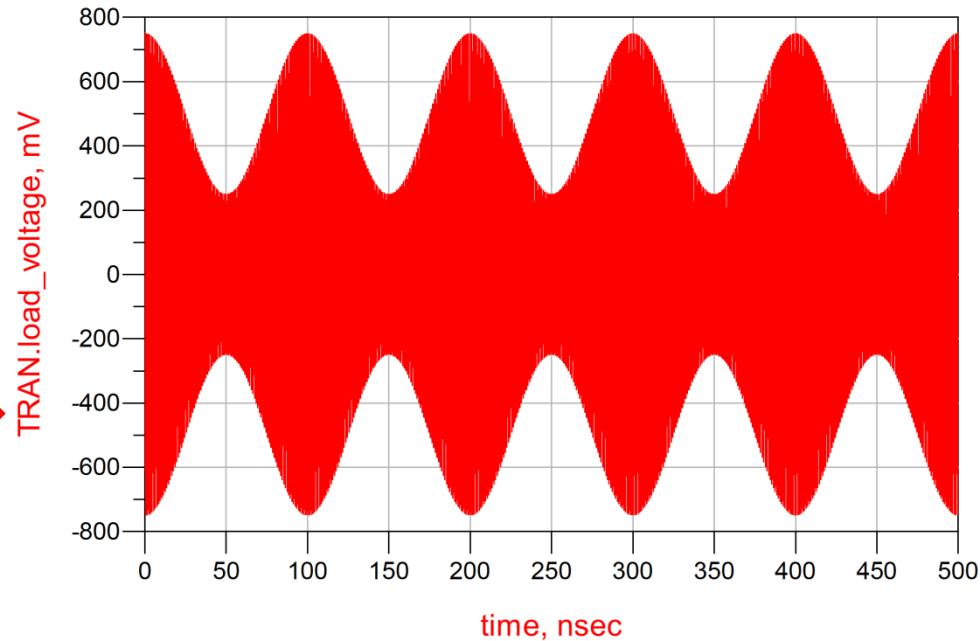
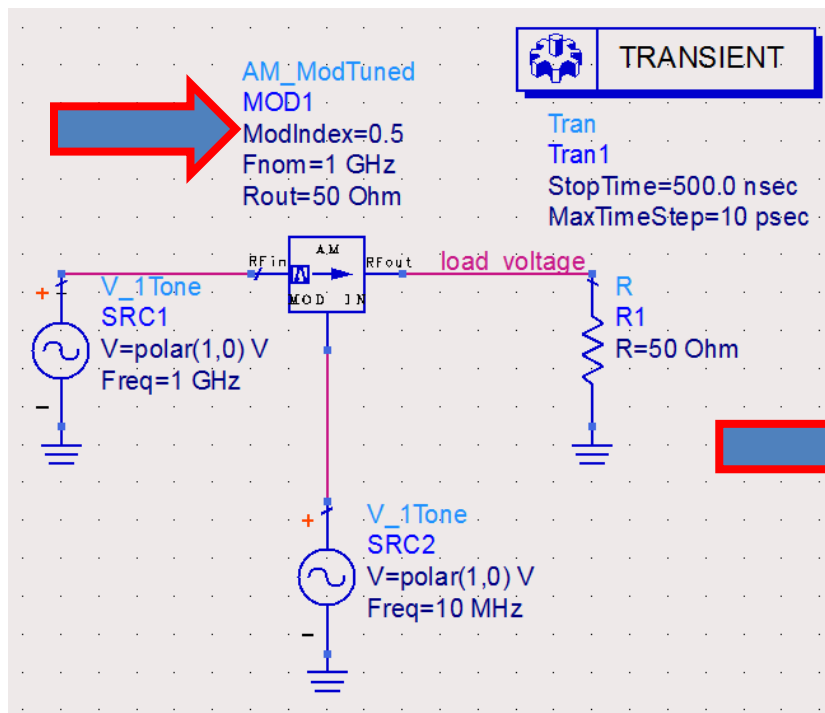
← Lower sideband
← Carrier
← Upper sideband

Notice that for $M=0.5$, the **sidebands** have a **QUARTER** of the **VOLTAGE** of the **carrier**. This means -12 dB!!!

Introduction to modulated signals



- Let's look at the simulation results:



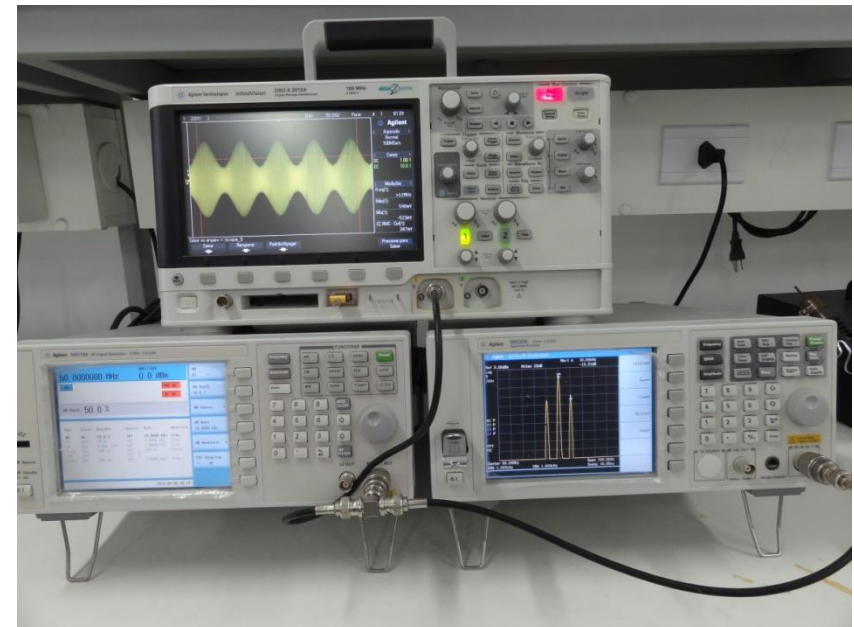
Introduction to measurements



- Let's check that with measurements:



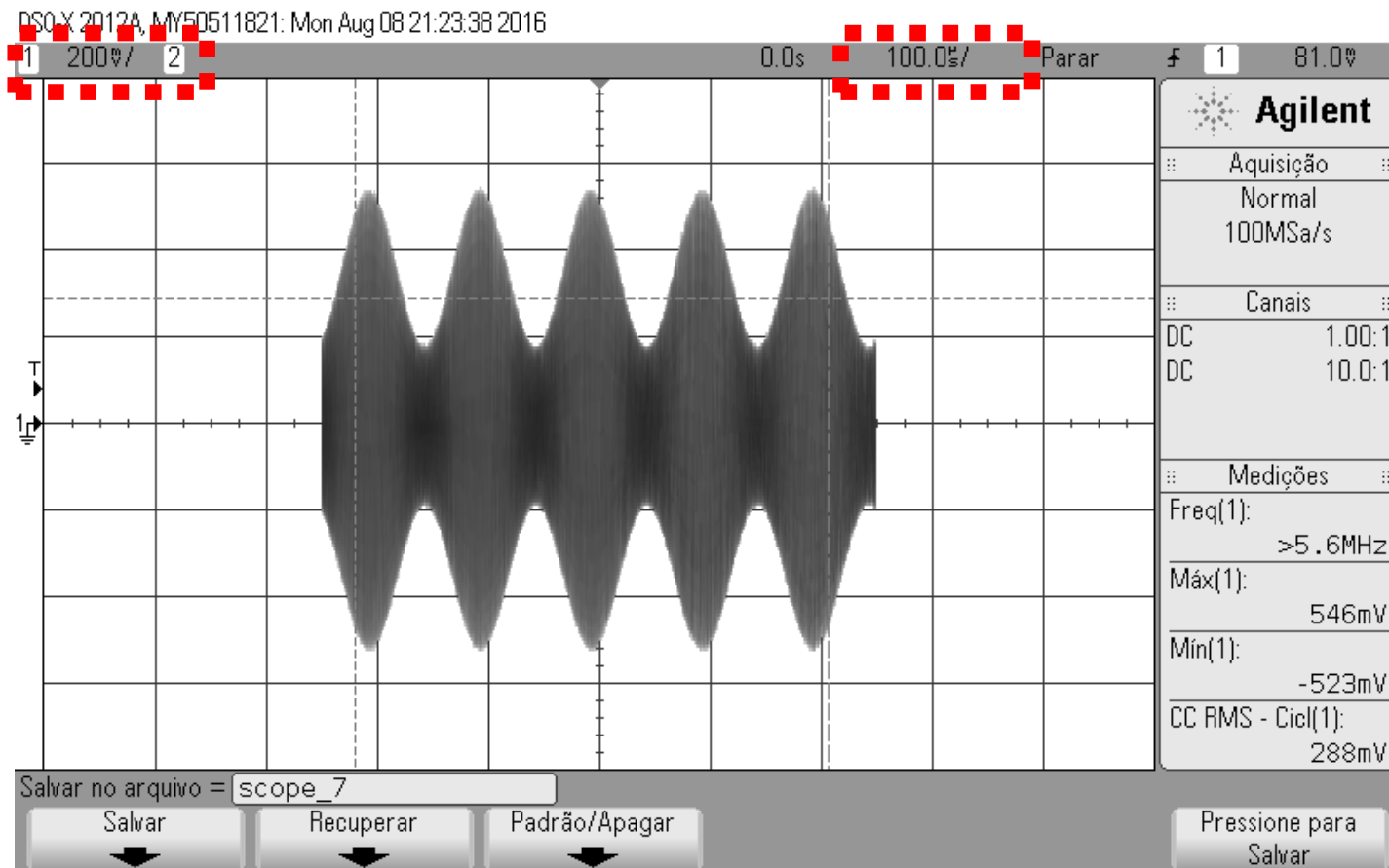
← M=0.5



Introduction to measurements



- Check the time-domain waveform:

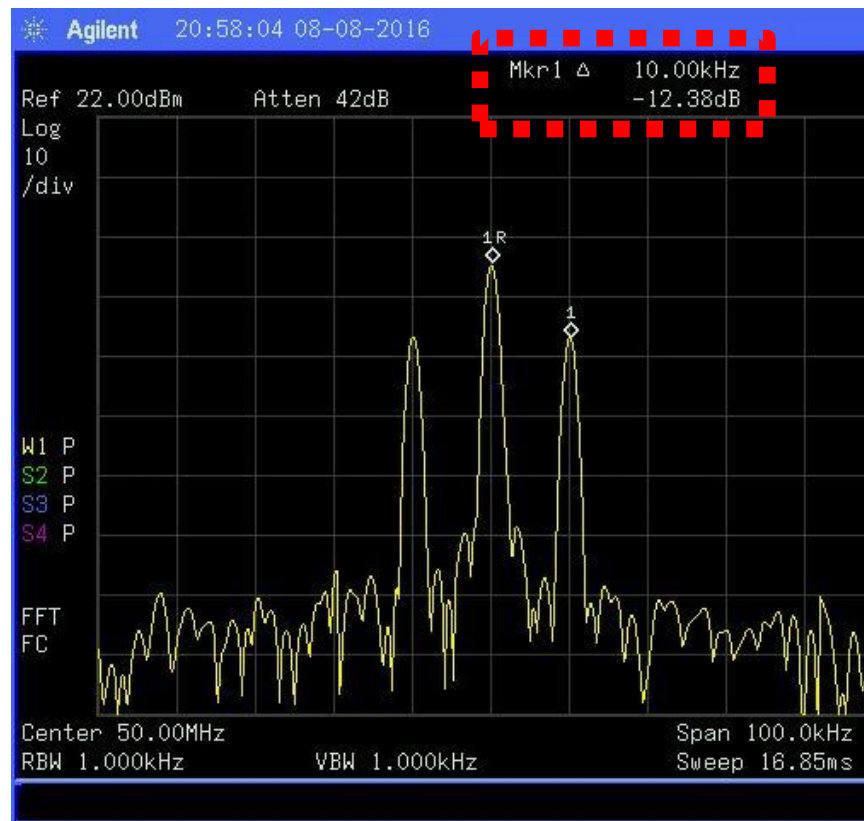


Signal truncated to ease measuring

Introduction to measurements



- Look at the frequency domain measurements:



Sounds good?

Introduction to modulated signals



- OK for AM, but what about the “marvelous” FM?
 - The mathematics is much more involved.
 - We will try to keep it simple (more on this later), and focus on the single-tone modulation:

$$FM_{SINGLETON}(t) = A_C \cos \left[\omega_c t + \frac{\Delta f}{f_m} \sin(\omega_m t) \right]$$

Maximum frequency deviation

Carrier amplitude

Carrier frequency

Modulating frequency

Introduction to modulated signals



- What about the spectrum (google it!)?

$$FM_{SINGLETON}(t) = A_C \cos \left[\omega_c t + \frac{\Delta f}{f_m} \sin(\omega_m t) \right]$$

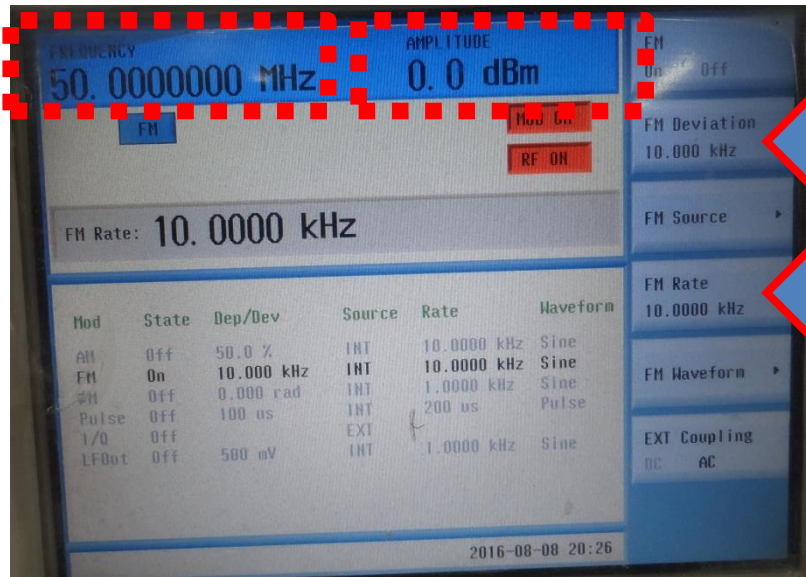
$$\frac{\Delta f}{f_m} = \text{Modulation index}$$

Modulation index	Sideband amplitude										
	Carrier	1	2	3	4	5	6	7	8	9	10
0.00	1.00										
0.25	0.98	0.12									
0.5	0.94	0.24	0.03								
1.0	0.77	0.44	0.11	0.02							
1.5	0.51	0.56	0.23	0.06	0.01						
2.0	0.22	0.58	0.35	0.13	0.03						
2.41	0	0.52	0.43	0.20	0.06	0.02					
2.5	-0.05	0.50	0.45	0.22	0.07	0.02	0.01				
3.0	-0.26	0.34	0.49	0.31	0.13	0.04	0.01				

Introduction to measurements



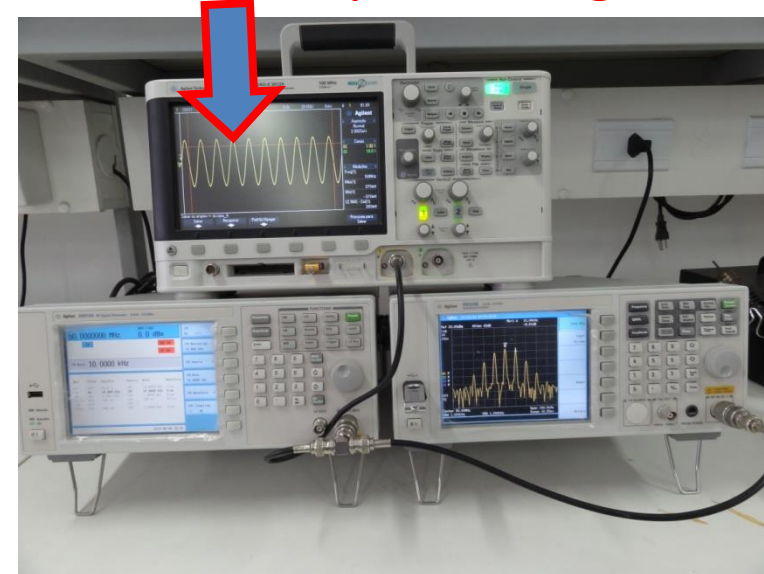
- Let's check that with measurements:



$\Delta f = 10 \text{ kHz}$

$f_m = 10 \text{ kHz}$

Constant amplitude signal!!

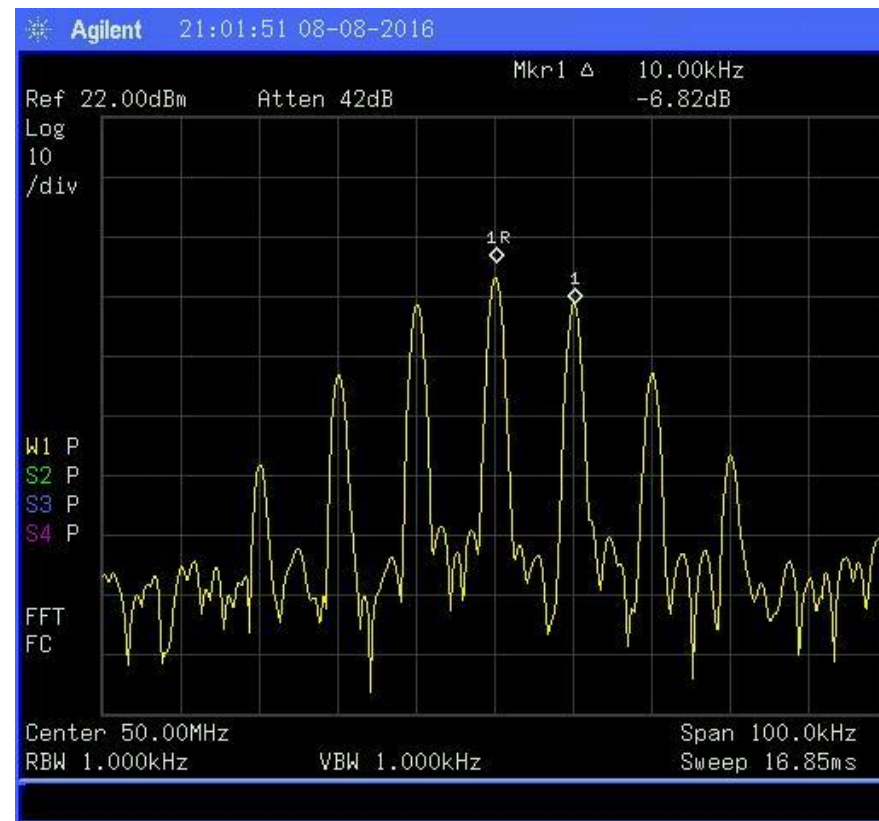


Introduction to measurements



- Look at the frequency domain measurements:

Modulation index					
	Carrier	1	2	3	4
0.00	1.00				
0.25	0.98	0.12			
0.5	0.94	0.24	0.03		
1.0	0.77	0.44	0.11	0.02	
1.5	0.51	0.56	0.23	0.06	0.01
2.0	0.22	0.58	0.35	0.13	0.03



Proposed exercises

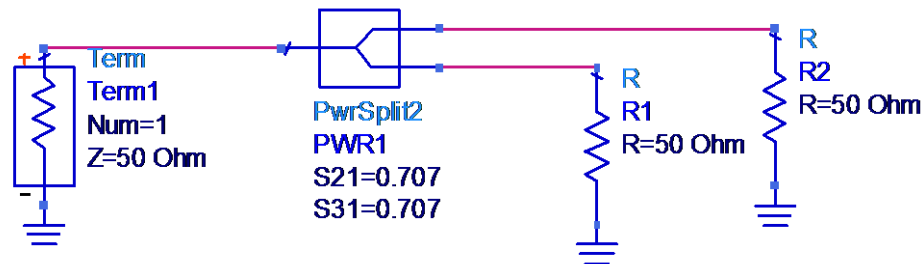


- Execute the following conversions:
 - 20dBm = _____ W
 - -40dBW = _____ dbm
 - $G=10\text{V/V}$ = _____ dB
 - $G=1\text{mA}/\mu\text{A}$ = _____ dB

Proposed exercises



A *Power Divider* (or *Power Splitter*) is a passive component which can be used to divide (split) the power of a source between two loads (supposedly the same). They are common in houses and buildings in which the same antenna (source) is divided between 2 television sets (loads). It can be imagined as a 3-port device (1 input, 2 outputs) which transform each of the two $50\ \Omega$ loads into a $100\ \Omega$, and put them in parallel, so the source “sees” a $50\ \Omega$ load. Consider the circuit below, composed by a signal source (“Term”), a *Power Divider* (“PwrSplit2”) and $50\ \Omega$ loads.



Supposing that the source of signal has an internal impedance of $50\ \Omega$ and available power of $-80\ \text{dBW}$, calculate the power, in dBm , in each of the loads (P_L).

Proposed exercises



- According to the Friss equation for transmission of signals, the ratio between the power received by an antenna (P_{RX}) and the transmitted power (P_{TX}) is:

$$\frac{P_{RX}}{P_{TX}} = G_{TX} \cdot G_{RX} \cdot \left(\frac{\lambda}{4\pi R}\right)^2$$

- In this equation, G_{TX} and G_{RX} indicate the TX and RX antenna gain respectively (with respect to the isotropic condition, in which power is evenly radiated in all directions), λ is the wavelength and R is the distance between the antennas. Supposing that a 2.4GHz transceiver is able to operate with a received power as low as -100 dBm and transmit 20dBm, if $G_{TX}=G_{RX}=0$ dB, what is the greatest distance between the antennas?

Proposed Experiments



- Change the frequency and amplitude of a single tone, and observe it on the oscilloscope and spectrum analyzer. Check values.
- Change the modulating frequency and modulation index of an AM modulated signal, and observe it on the oscilloscope and spectrum analyzer. Check values.
- Change the modulating frequency and frequency deviation of a FM modulated signal, and observe it on the spectrum analyzer. Check values.